Blind $L1$-norm Multichannel Deconvolution for Seismic Source Function Estimation

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Summary
In this work, similar to that proposed by Canadas (2002), synthetic data were used to analyze a blind deconvolution method to estimate source wavelets, which consists in alternating with single-channel reflectivity estimation, and multichannel wavelet estimation. The method is able to estimate the correct source wavelet under a wide range of conditions such as sparsity in the reflectivity series and band-width in the wavelet, and noise in the traces without making any a priori assumption of its phase.

Introduction
A seismic signal $s_t$ can be represented as

$$s_t = w_t * q_t + n_t$$  \hspace{1cm} (1)

where * means convolution, $w_t$ is the seismic wavelet, $q_t$ is the reflectivity and $n_t$ is the additive noise. The goal of deconvolution is, given $s_t$, to recover either $q_t$ or $w_t$.

Conventional deconvolution methods depend on the previous knowledge of the properties of both reflectivity and wavelet; however, in this case we will not assume a previous knowledge about the wavelet phase (blind deconvolution).

Theory
Estimation of the wavelet given the reflectivity
Estimation of the wavelet when the reflectivity is known reduces to a multi-channel damped least-squares solution with quadratic regularization. The cost function from equation (1) is given by

$$J_w = \frac{1}{N} \sum_{j=1}^{n} \left\| Q_j w - s_j \right\|_2^2 + \mu \left\| w \right\|_2^2$$  \hspace{1cm} (2)

Where $\mu$ is the damping parameter and $j$ is the number of traces. Minimizing equation (2) with respect to the unknown wavelet and equating it to zero we have that

$$w = \left( R_q + \mu I \right)^{-1} g_q$$  \hspace{1cm} (3)

where
\[ R_q = \frac{1}{N} \sum_{j=1}^{N} Q_j^T Q_j \]  

is the average autocorrelation matrix of the reflectivities, and

\[ g_q = \frac{1}{N} \sum_{j=1}^{N} Q_j^T s_j \]  

is the average cross-correlation vector of the traces with the reflectivities.

**Estimation of the reflectivity given the wavelet**

Estimation of the reflectivity when the wavelet is known involves defining as many cost functions as traces. Thus, from equation (1) we have that

\[ J_{q_j} = \| Wq_j - s_j \|_2^2 + \beta R(q_j), \quad j = 1, \ldots, N. \]  

In order to recover a sparse reflectivity sequence, the non-quadratic regularization term \( R(q) = \sum_t |q_t| \) and the associate hyperparameter \( \beta \) have been introduced in the cost functions.

\[ \sum_t |q_t| \] is known as the \( L_1 \)-norm of \( q \) and has been studied by different authors (Taylor et al., 1979; Levy and Fullegar, 1981; Oldenburg et al., 1983; Santosa and Symes 1984). Minimizing equation (6) with respect to the unknown wavelet and equating it to zero we have that

\[ q_j = \left( W^T W + \beta \Lambda(q_j) \right)^{-1} W^T s_j, \quad j = 1, \ldots, N. \]  

where

\[ \Lambda(q)_{kk} = \frac{1}{|q_k| + \epsilon} \]  

being \( \epsilon \) an stabilization term.

Because equation (7) is non-linear, its solution must be found using an iterative approach like Iteratively Reweighted Least Squares (IRLS):

\[ q_j^{(k+1)} = \left( W^T W + \beta \Lambda(q_j^{(k)}) \right)^{-1} W^T s_j, \quad j = 1, \ldots, N. \]  

**Examples**

The code was implemented statistically in a wide range of synthetic data set in order to test its stability. Each data set was generated by convolving a particular wavelet with a particular reflectivity. Features of both wavelets and reflectivities used are described below:

**Wavelets**: trapezoidal amplitude spectrum wavelets with corner frequencies given by \( f=[f_1,f_2,f_3,f_4] \) were used. Frequencies \( f_1 \) and \( f_2 \) remained constant at \( f_1=1 \) and \( f_2=2 \) Hz while \( f_3 \) and \( f_4 \) were varied in the values of [20,30,40,60] and [30,40,50,70] [Hz] respectively, where \( f_3 < f_4 \). In each combination of \( f_3 \) and \( f_4 \), a constant phase of \( c=[0,10,20,30,45,60] \) [deg] was applied.
Reflectivity: The code is very sensible to the number of traces that is chosen. After experiencing with different N number of traces, convergence is achieved for N > 10. In this work, data set consisted of 20 uncorrelated traces of n=100 samples per trace. The amplitude of the reflector has a non-Gaussian distribution with sparsity δ varying in a range of [0.1, 0.3, 0.6, 0.8]. Sparsity defines the fraction of non-zeros reflectors in the reflectivity series.

Data was contaminated with white Gaussian noise with both signal to noise ratio of 5 and 20. Kurtosis was used to calculate the phase correction that was applied to the estimated wavelet. The maximum of the Kurtosis indicates the correct phase correction needed to convert the estimated wavelet in a zero phase wavelet. Figures 1 and 2 show examples of the performance of the algorithm under different values of SNR, sparsity and frequency.

![Figure 1](image1.png)

Figure 1. Example of the performance of the code when SNR=20, sparsity=0.8, and constant phase added of c=30: (a) true wavelet (left) and estimated wavelet (right), (b) kurtosis vs phase correction. Maximum of the kurtosis corresponds to a phase correction of -33 deg.

![Figure 2](image2.png)

Figure 2. Example of the performance of the code when SNR=5, sparsity=0.1, and constant phase added of c=30: (a) true wavelet (left) and estimated wavelet (right), (b) kurtosis vs phase correction. Maximum of the kurtosis corresponds to a phase correction of -29 deg.

Conclusions
The assumption of non-correlated traces on which this code was developed is fundamental to guarantee the adequate performance of this method. The quality of the result also depends on choosing sufficient number of seismic traces. The level of sparsity of the reflectivity series and the amplitude content of the seismic wavelet do not affect phase estimations. In summary, this code has the potential to be used in real data that shows a complex geology.

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**References**


